



E-ISSN: 2706-8927
P-ISSN: 2706-8919
www.allstudyjournal.com
IJAAS 2020; 2(3): 734-737
Received: 26-03-2020
Accepted: 24-04-2020

Dr. Ranjeet Kumar
+2 Teacher, BP Inter School,
Begusarai, Bihar, India

A study of low frequency RFID system

Dr. Ranjeet Kumar

Abstract

In this paper, a functioning bandpass channel is required for the front-end framework since all signs outside the (10-20) kHz time-shifting sign backscattered by the label should be dismissed. The channel is developed from some given determinations, one of which is, that the channel needs to have a Butterworth reaction. The engineering that will be utilized is the Sallen-Key.

Keywords: RFID system and Sallen-Key

Introduction

Radio frequency distinguishing proof (RFID) framework is a remote correspondence framework that is utilized to recognize labeled articles, individuals or creatures. The zone of uses for RFID is expanding quickly. Applications include flexibly chain the executives, access control to building, security frameworks, creature recognizable proof, public transportation, medical services, open-air occasions, air terminal things, abundance bundle coordination's etc.

RFID framework comprises of perusers and an enormous number of labels. A tag has a distinguishing proof number (ID) and a peruser perceives an article through back to back correspondences with the label connected to it. The peruser conveys a sign which supplies force and guidelines to a tag. The tag sends its ID to the peruser and the peruser counsels an outer information base with got ID to perceive the article. In this paper, RFID framework is considered with 125 kHz, FSK balance conspire.

In the peruser, the front-end framework needs a RC channel, a functioning band-pass channel and a functioning low-pass channel to dismiss the undesired signs.

Channels are basic parts in numerous electrical frameworks. In best in class RF recipients, elite channels are needed to eliminate undesired signs at various phases of the getting cycle, for example, commotion from approaching signs the receiving wire gets, undesired signs at the picture recurrence, and sounds after the blending activity. All simple channels fall in one of two classes: uninvolved or dynamic. In this low recurrence RFID framework, dynamic channels are utilized due to the accompanying points of interest:

- Active filter can generate a gain larger than one.
- Higher order filters can easily be cascaded since each Opamp can be second order.
- Filters are smaller in size as long as no inductors are used, which makes it very useful as integrated circuits.

In this paper, a functioning band-pass channel is planned and recreated. A functioning band-pass channel is utilized for the RFID framework to dismiss all signs outside the (10-20) kHz signals and to enhance the low reception apparatus signal. These are on the grounds that the ID signals from the tag are 12.5 kHz and 15.65 kHz and sign force is extremely low.

The most widely recognized channel reactions are the Butterworth, Chebyshev, and Bessel types. Among these reactions, Butterworth type is utilized to get a maximally-level reaction. Additionally, it shows an almost level pass band with no wave. The move off is smooth and monotonic, with a low-pass or highpass move off 20dB/dec for each post. Hence, a fourth request Butterworth band-pass channel would have a constriction pace of - 40dB/dec and 40 dB/dec.

In the second segment of this paper, the band-pass channel configuration will be resolved with its determinations. And afterward execution of fourth request Butterworth band-pass channel configuration will be completed so as to meet the plan determinations. The third and last aspect of this paper, the examination of the Circuit Maker reenactment result and MATLAB recreation result will be talked about.

Corresponding Author:
Dr. Ranjeet Kumar
+2 Teacher, BP Inter School,
Begusarai, Bihar, India

Design consideration

The design that has been utilized to execute the fourth request band-pass channel is Sallen-Key Topology. This was picked as a result of its straightforwardness contrasted with other referred to models, for example, multiple input and state variable, where the last is for exactness performance. Butterworth channel reaction is utilized to get the greatest level increase. The Active - RC Butterworth channels have a scope of points of interest when utilized for lower request of the channel: have phenomenal linearity, have low force dissemination and are anything but difficult to plan and investigate. The channel reaction is unfeeling toward parasitic, and it has huge Dynamic reach.

A circuit graph for second request Sallen-Key band-pass channel is appeared in Fig 1.

Table.1 represents the details for the ideal bandpass channel. By utilizing the accompanying channel boundaries, the necessary channel can be planned and reenacted with circuit creator and MATLAB.

Design implementation

The transfer function for the fourth order band-pass filter is: where

$$A(s) = \frac{\frac{A_{mi}(\alpha, s)}{Q_i}}{1 + \frac{\alpha \cdot s}{Q_i} + (\alpha \cdot s)^2} \cdot \frac{\frac{A_{mi}(\alpha, s)}{Q_i}}{1 + \frac{1}{Q_i} \left(\frac{s}{\alpha} + \frac{1}{\alpha} \right)^2} \tag{1}$$

- A_{mi} is the gain at the mid frequency, f_{mi} , of each filter.
- Q_i is the pole quality of each filter.
- α and $1/\alpha$ are the factors by which the mid frequencies of the individual filters, f_{m1} and f_{m2} , are derived from the mid frequency, f_m , of the overall band-pass. Factor α needs to be determined through successive approximation, using the following equation.

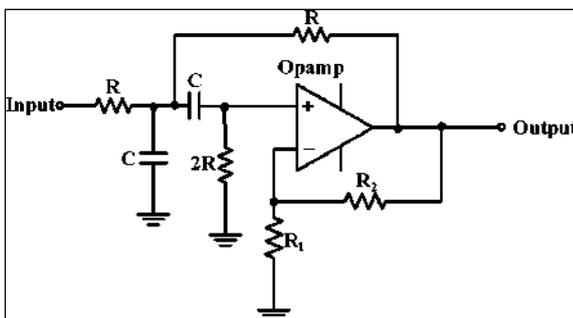


Fig 1: General architecture of a second order Sallen-Key filter

Table 1: Band-pass Filter Specifications

Center frequency	15 kHz	
Pass band Frequency	10 kHz	20 kHz
Pass band Ripple	0.1dB	
Stop band Attenuation	60dB	
V_{pp}	100μV	

$$\alpha^2 + \left[\frac{\alpha \cdot \Delta\Omega \cdot a_1}{b_1 (1 + \alpha^2)} \right]^2 + \frac{1}{\alpha^2} - 2 - \frac{(\Delta\Omega)^2}{b_1} = 0 \tag{2}$$

where normalized bandwidth $\Delta\Omega=1/Q_{BP}$, (Q_{BP} is the overall quality of the filter), with a_1 and b_1 being the second order low-pass coefficients of the desired filter type.

For Butterworth filter type,

- $a_1 = 1.4142$ and
- $b_1 = 1$

Fourth-order Butterworth Band-Pass filter with the following parameters:

Passband frequencies = (10-20) kHz

Mid-frequency of the overall filter,
 $f_m = 15$ kHz

Let overall gain at mid-frequency,

$$A^m = 2$$

Bandwidth, $BW = f_H - f_L = 10$ kHz

$$Q_{BP} = \frac{f_m}{BW} = 1.5$$

By using equation (2),

$\alpha = 1.2711$ is obtained.

After α has been determined, all quantities of the partial filters can be calculated as follows:

The mid frequency of filter 1 is:

$$f_{m1} = \frac{f_m}{\alpha} = 11.8 \text{ kHz}$$

The mid frequency of filter 2 is:

$$f_{m2} = f_m \cdot \alpha = 19.067 \text{ kHz}$$

The individual pole quality, Q_i , is the same for both filters:

$$Q_i = Q_{BP} \cdot \frac{(1 + \alpha^2) b_1}{\alpha \cdot a_1} = 2.1827$$

The individual gain (A_{mi}) at the partial mid- frequencies, f_{m1} and f_{m2} , is the same for both filters:

$$A_{mi} = \frac{Q_i}{Q_{BP}} \cdot \sqrt{\frac{A_m}{b_1}} = 2.0579$$

For Sallen-Key architecture, the required parameters can be calculated by the following equations:

$$\text{mid-frequency: } f_{mi} = \frac{1}{2\pi RC}$$

$$\text{inner gain: } G = 1 + \frac{R_2}{R_1}$$

$$\text{gain at } f_m: A_{mi} = \frac{G}{3 - G}$$

$$\text{filter quality: } Q_i = \frac{1}{3 - G}$$

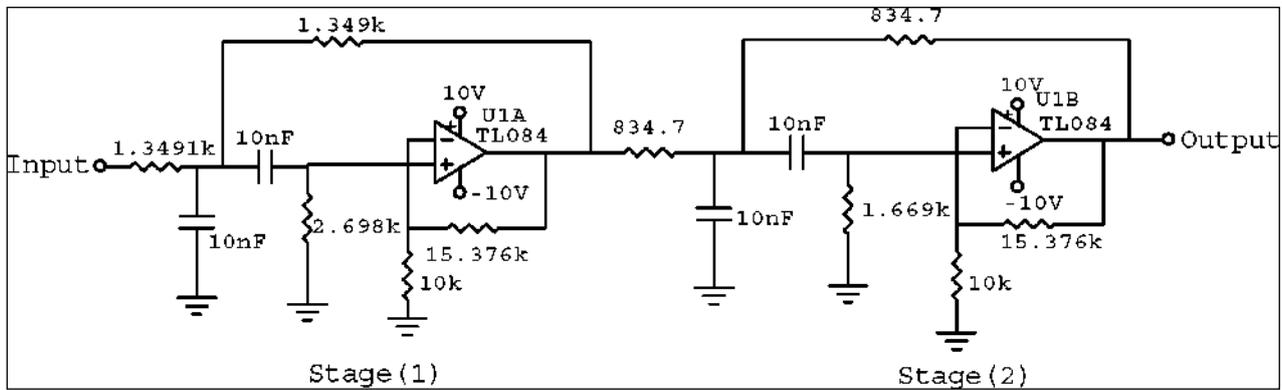


Fig 2: Schematic scheme of 4th order Butterworth Band-pass filter

To design the individual second-order band-pass filters, specifies C = 10 nF, the resistor values for both partial filters are calculated as mentioned below:

Filter 1:

- R = 1.3491 kΩ
- Let R₁ = 10kΩ
- R₂ = 15.376 kΩ

Filter 2:

- R = 834.7 Ω

Let

- R₁ = 10kΩ
- R₂ = 15.376 kΩ

The 4th order Butterworth band-pass filter is constructed from two non-identical 2nd-order sections shown in Fig.2. The transfer function of that circuit is

$$H(s) = \frac{0.8889s^2}{s^4 + 0.9427s^3 + 3.4413s^2 + 0.9428s + 1}$$

By using the transfer function, the frequency response of the filter can be plotted using MATLAB to verify the design.

Results

The consequences of circuit producer for the fourth request Active-RC Butterworth band-pass channel are appeared in Fig. 3. Fourth request Active - RC Butterworth channel configuration has passband frequencies 10 kHz and 20 kHz, passband addition of around 49 dBV and move off paces of - 40dB/dec and 40dB/dec.

Likewise, outlines the recurrence reaction of the channel utilizing MATLAB reproduction technique. It very well may be seen that the reenacted reaction looks great and furthermore looks acquainted with the reproduced reaction in MATLAB, and in this way chose to be actualized in reality.

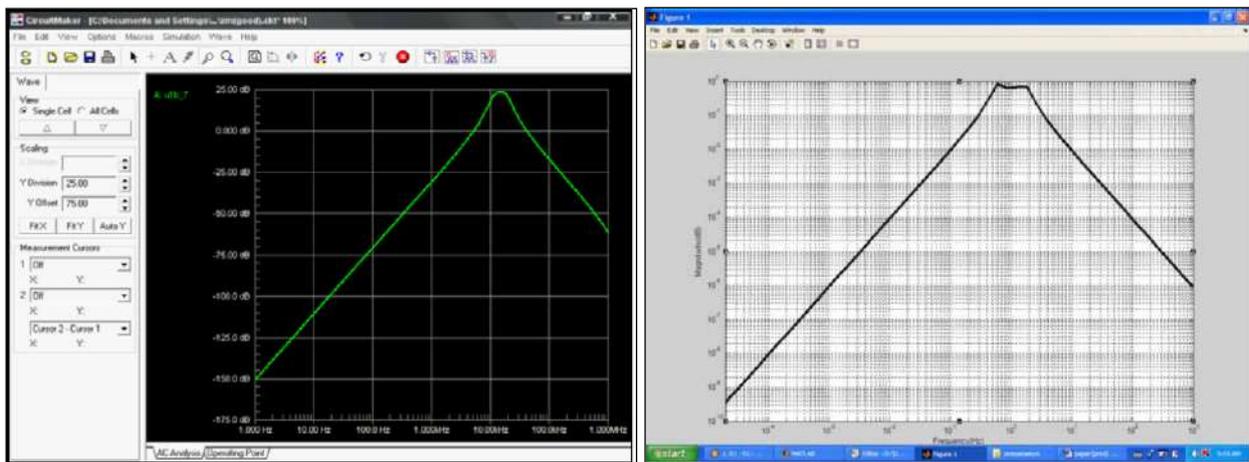


Fig 3: Simulated filter responses in circuit maker of fourth order Butterworth Band-pass filter

Conclusion

In this paper these circuits are made out of two Op-amps, ten resistors, and four capacitors. The speakers depend on the TL084 circuits as the increase component since this circuit is useful for this RFID application. As the recreated outcomes fulfill the framework prerequisites, these circuit structures are reasonable for RFID application. In the event that more-exact recurrence reaction is required, more stages ought to be utilized.

References

1. Löwenborg P, Gustation O, Wanhammar L. Filter Design Using MATLAB, Radiovetenskap och

- Kommunikation 99, Karlskrona, June 1999.
2. Löwenborg P, Johansson H, Wanhammar L. "A Class of Two- Channel Hybrid Analog/ Digital Filter Banks", IEEE Midwest Symp. Circuits Systems, Las Cruces, New Mexico, Aug 1999.
3. Löwenborg P, Johansson H, Wanhammar L. "A Design Procedure for 2-Channel Mixed Analog and Digital Filter Banks for A/D Conversion Using Minimax Optimization", IEEE Int. Conf. Elect. Circuits Syst., Cyprus, Sept 1999.
4. Wanhammar L, Johansson H. "Analog filter", Linköping University, 300 pages 2000.

5. Larry D Paarmann. "Design and analysis of analog filters", A Signal Processing perspective, Wichita State University, 436 pages 2003.
6. Ron Weinstein. "RFID: a technical overview and its applications to enterprise", IEEE Computer Society, June 2005.
7. "125 kHz micro IDTM Passive RFID device". [Online]. Available: <http://www.microchip.com>.
8. Winder S. "Analog and Digital Filter Design", 2nd ed. Woburn, MA: Newnes 2002.
9. "TLO84 data sheet", Texas Instruments, [Online] 1999. Available: <http://www-s.ti.com/sc/ds/tlo84.pdf>.
10. "FSK Reader Reference Design", DS51137B, Microchip Technology Inc 1998. Available: <http://www.microchip.com>.